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Author(s)	Kambe, Kazuko; Iwagaki, Tsuyoshi; Inoue, Michiko; Fujiwara, Hideo
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Description	

# Efficient Constraint Extraction for Template-Based Processor Self-Test Generation

Kazuko Kambe<sup>†</sup>    Tsuyoshi Iwagaki<sup>‡</sup>    Michiko Inoue<sup>†</sup>    Hideo Fujiwara<sup>†</sup>

<sup>†</sup>*Graduate School of Information Science, Nara Institute of Science and Technology  
Kansai Science City 630-0192, Japan*

<sup>‡</sup>*School of Information Science, Japan Advanced Institute of Science and Technology  
1-1 Asahidai, Nomi, Ishikawa 923-1292, Japan*

E-mail: <sup>†</sup>{kazuk-ka, kounoe, fujiwara}@is.naist.jp, <sup>‡</sup>iwagaki@jaist.ac.jp

## Abstract

*This paper presents efficient method to extract constraints from a test program template and synthesize a test program using constraint circuits. A test program template is an instruction sequence with unspecified operands, and represents paths for justification of test patterns and observation of test responses for a module under test (MUT). The constraint circuit represents a relation between operand values and inputs/output of the MUT, therefore it enables to obtain operand values using a standard automatic test pattern generator. Experimental results show that the proposed method generates accurate and compact constraint circuits, and we obtain high fault efficiency.*

## 1. Introduction

Recent processor designs with giga-hertz speed and rich functionality demand for accurate and at-speed testing. Software-based self-test attracts attention as a testing strategy enabling at-speed test [1, 2, 3, 4, 5, 6, 7]. In this approach, a processor enables self-test by running a sequence of instructions called a test program. Therefore, it does not induce any performance penalty, area overhead or excessive power. Recent works are towards hierarchical test generation approach targeting structural faults where gate level tests for modules in processors are linked with test programs at processor level. In this approach, values or sequences of values applicable to the modules inside the processor are restricted as a test program delivers test patterns to the modules. These works extract such restrictions for a module under test (MUT) as constraints, then apply an automatic test pattern generator (ATPG) to the MUT with the extracted constraints. Finally, the generated tests are transformed into test programs.

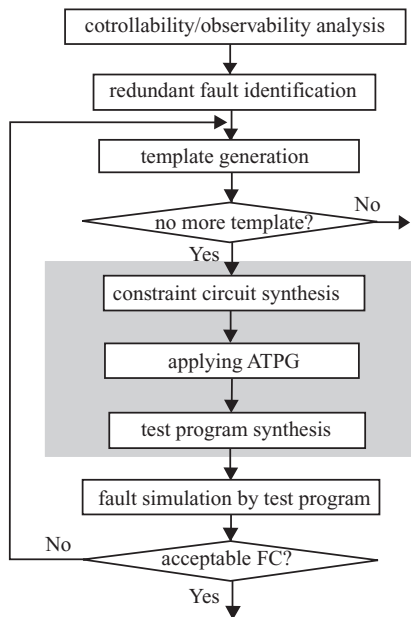
The method in [8] reduces a target circuit to a subset of the circuit surrounding an MUT by analyzing RTL descriptions and applies ATPG to the subset circuit. However, the

generated test patterns might induce overtesting at instruction level, since the method does not consider constraints imposed instructions. Furthermore, it is possible that the extracted subset is the same size as the original design in the worst case.

The methods in [5], [6] and [7] are proposing test program generation based on test program templates (or just templates) targeting stuck-at faults. A template is an instruction sequence with unspecified operands that delivers tests to an MUT and observes the test responses. The methods extract constraints from each template since the template represents ways to propagate tests from primary inputs and the test response to primary outputs. Therefore, the generated tests for a module are guaranteed to be transformed into test programs. However, in [5], constraints are obtained by a regressive analysis for randomly generated values of operands, and it might be difficult to find accurate constraints if mapping functions between operands and MUT inputs are complicated.

In this paper, we present an efficient method to extract constraints from a template that generated by our previous method in [6]. Our technique analyzes RTL descriptions of a processor according to behavior of a template and synthesizes constraint circuits, which work as mapping functions between operands and MUT inputs. We obtain operand values from generated test patterns using standard ATPG. The method has the following advantages: (i) we can generate an accurate and compact constraint circuits even if a relation between operands and MUT inputs is complicated, (ii) a constraint circuit is synthesized as a combinational circuit by applying time expansion to a template, thereby we apply combinational ATPG to a combinational MUT.

This paper is organized as follows. In Sections 2 and 3, we give a brief explanation of our self testing system and template generation method in [6]. A method of constraint extraction is proposed in Section 4, and the experimental results are shown in Section 5. Finally, Section 6 concludes this paper.



**Figure 1. Overview of test program generation.**

## 2. Template based test program generation

In our methodology, constraints are extracted by templates. A template is a sequence of instructions with unspecified operands, and represents paths for justification of test patterns and observation of test responses for an MUT.

**Example 1** An example in Figure 2 is a template for testing Arithmetic Logic Unit (*ALU*) of Parwan processor [9], where *loc1*, *loc2*, *loc3* and *loc4* are operands of instructions. The values are unspecified. First four instructions, CLA, ASL, LDA and ADD, justify values of Status Register (*SR*) and Accumulator (*AC*). Next SUB applies test patterns to *ALU*. STA observes the test response stored in *AC*. □

We explain an overview of our test program generation in Figure 1. We first analyze controllability and observability of registers to use for instruction selection. We identify redundant faults prior to test program generation. Here, we consider a fault is redundant if no test program detects the faults. The method in [7] proposed to identify redundant faults for sequential modules. Control signals applicable to combinational modules from a controller are restricted to constant values imposed by instruction code. We apply such restriction to the MUT as constraints. For each MUT, we repeat the following. We generate a template, and synthesize input/output constraints of the MUT from the template. After applying ATPG, a test program is transformed from the test patterns. In the next step, a fault simulation is applied to an entire processor. We generate multiple templates to cover input space for MUTs. These processes are repeated until all templates are generated or fault coverage reaches the re-

quired level. A methodology for template generation was described in [6]. In this paper, we focus the steps from constraint circuit synthesis to test program synthesis in shaded area of the figure.

## 3. Template generation

In our template generation method, we first select instructions to justify values of each input adjacent registers. Input/output adjacent registers of an MUT are defined as registers that are connected with inputs/outputs of the MUT directly or indirectly through combinational circuits. Once values of input adjacent registers are justified, the test patterns can be applied to inputs of the MUT. If some registers still require justification, more instructions should be selected until source registers of instructions reach the memory. We also select instructions to propagate the fault effect to the memory such that the instructions observe a value of an output adjacent register, and append to the end of the template. Since one template may not cover the whole input space of an MUT, we generate more templates. Our method uses an IE-Graph proposed in [4] to explore delivering routes for justification and propagation. An IE-Graph is a model of data flow between registers. Nodes are registers, IN and OUT, which model external world such as the memory. A directed edge between two nodes represents data transfer between registers controlled by an instruction. More details regarding template generation method can be found in [6].

## 4. Constraint extraction

Figure 2 illustrates an outline of test program generation for a template. The *ALU* has four inputs (*a\_side*, *b\_side*, *alu\_code* and *in\_flags*) and two outputs (*alu\_out* and *alu\_flags*). We synthesize input/output constraints from the template. The circuits describe relations between values of operands (*mem[loc1]*, *mem[loc2]*, *mem[loc3]* and *mem[loc4]*) and inputs/output of *ALU* in HDL coding style. We apply ATPG to a circuit that consists of input/output constraint circuits and the MUT, targeting faults in the MUT. Now, generated test patterns are directly linked to the operand values. In consequence, it is easy to synthesize a test program.

### Time expansion of template

We are interested in relations between values of operands and inputs/output of an MUT. In our method, we perform time expansion of a template, and eliminate registers. We represent necessary data flow for justification/observation paths as a sequence of module instances where modules have different instances for different time frames. We can naturally obtain such data flow during traversing paths in an IE-graph to select instructions for template generation. We construct an instance data flow graph (IDFG) to represent data flow between module instances. Figure 3 shows an example of a partial IDFG. It corresponds to a part of

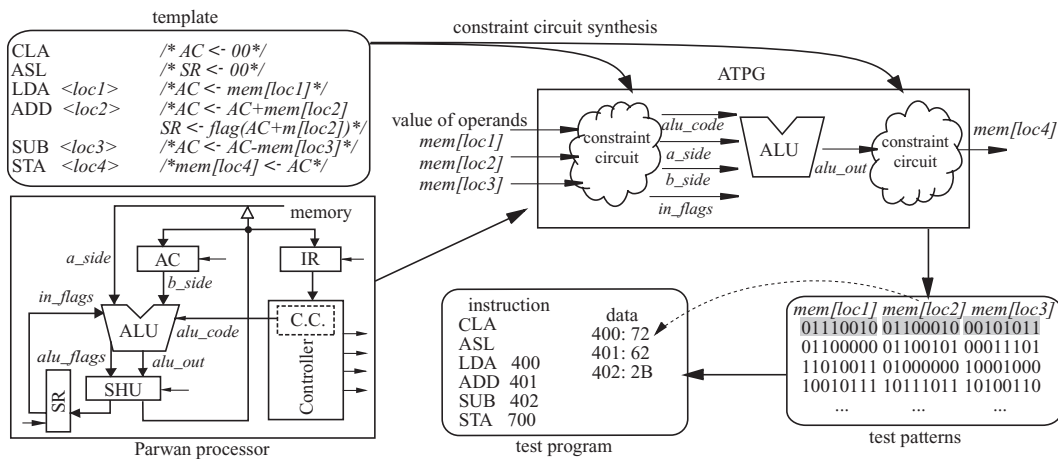


Figure 2. Test program generation.

execution of an instruction sequence: ADD, SUB and STA. Nodes of an IDFG are instances of modules, IN or OUT. Each node is marked with a [state, instruction] pair, which means that data transfer occurs at the state of a controller during executing the instruction where an execution of the instruction consists of multiple cycles. Each module instance has a RTL description duplicated from the description of the module. A directed edge between represents data flow between nodes. Each edge is labeled with a set of [source signal name, destination signal name] pairs. The signal names are described in RTL descriptions. This information of labels is used to simplify RTL descriptions of modules.

### Optimization and synthesis of constraint circuits

When a module is instantiated to a specified state of controller during the execution of an instruction, all of the paths in modules are not activated. Therefore, extracting meaningful functions of modules regarding a specified instruction enables to reduce RTL descriptions. We will describe how to optimize RTL descriptions.

In a RTL description of module behavior, conditional branch statements are generally used to specify a set of statements to execute based on value of given selection signal. Such signals might be given from a controller. Analyzing behavioral descriptions of a controller, we can find that control signals have constant values corresponding to

each state during executing an instruction. The following fragment of VHDL code is a RTL description of ALU in Parwan processor.

```

case alu_code is
  when a_input =>
    t := "00" & a_side;
    c := in_flags(2); v := in_flags(3);
  when ...
    ...
  when others =>
    t := "0000000000";
    c := '0'; v := '0';
end case;

```

When the processor executes LDA, a signal value *alu\_code* is "a\_input" for a state, and 3rd and 4th lines are executed. Other statements are unnecessary. Consequently, we easily extract functions of modules which are instantiated during executing an instruction, and eliminate dead codes and unnecessary ports once values of signals are specified to constant values.

Now, we synthesize constraint circuits in HDL code from an IDFG. We first apply topological sort to nodes of the IDFG, in order to divide into two parts, justification and observation. After applying topological sort, the total ordered node sequence is obtained. For each module instance in the ordered sequence, we generate behavioral descriptions by optimizing in ways described above. We also describe structures in HDL codes for input/output constraint circuits, where module instances are connected together with signals according to data flow dependence.

For sequential modules, an *input temporal spatial constraint* was introduced in [7]. An input temporal spatial constraint of a module is a sequence of input constraints. We extract a constraint for each cycle from the first cycle of the instruction that applies test patterns to an MUT, to the instruction fetch cycle of the next instruction in a template. An input constraint circuit for a sequential MUT consists of a simple sequencer and some combinational constraint circuits where a combinational constraint circuit represents

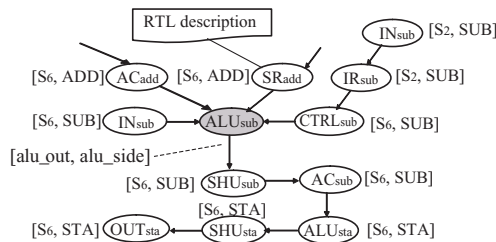


Figure 3. Instance data flow graph.

**Table 1. Area of constraint circuits.**

	ALU	SHU	CTRL	PC
module	460	140	407	468
input constraint	221.0	185.0	792.0	746.5
output constraint	0.0	0.0	-	0.0

The area of Parwan processor : 2800

**Table 2. Result on Parwan modules.**

	ALU	SHU	CTRL	PC
#generated templates	276	276	299	559
#contributive templates	12	11	19	3
#total faults	886	264	666	544
#faults detected by ATPG	881	258	622	436
#redundant faults	2	1	29	0
#faults detected by fault simulation	863	257	596	540
#fault efficiency(%)	97.63	97.73	94.14	99.26
test program length	564	228	74	253

a constraint for a cycle and the sequencer selects the constraint corresponding to the cycle.

## 5. Experimental results

Extracting meaningful functions of module instances yields dead codes and unnecessary ports. In this paper, we eliminated such dead codes and ports, and synthesized constraint circuits by a logic synthesis tool Design Compiler from Synopsys.

We evaluated the proposed method for Parwan processor. Table 1 shows the average of constraint circuit area for modules. The area of output constraint circuits for *ALU*, *Shifter Unit (SHU)* and *Program Counter (PC)* are 0.0, however, the circuits have some lines. This shows that the value of MUT outputs can be observed at primary outputs as it is. In this experiment, we assume a fault to be detectable if an error is propagated to the outside of a controller; accordingly, output constraint circuits for *Controller (CTRL)* were not synthesized. Notice that every constraint circuit is small. The input constraint circuits of *CTRL* and *PC* which are sequential modules are more than 1.5 times of the modules. However, the average of constraint circuit area per cycle is 126.2 for *CTRL*, and 98.2 for *PC*, respectively. The constraint circuits for sequential modules are also compact.

We combined constraint circuits with an MUT, and applied ATPG tool TestGen from Synopsys to the circuit. The results are shown in Table 2. Row 2 is the total number of generated templates, and row 3 is the number of templates that actually contributed to detecting faults. Other rows show total faults of each module, the number of faults detected by ATPG. Test programs detected faults in row 7. Consequently, we can achieve high fault efficiency for every module. The experimental results show that our method can synthesize accurate and compact constraint circuits efficiently.

## 6. Conclusion

This paper presented an efficient constraint extraction method based on templates. We proposed systematic method to extract constraints and synthesize test programs using constraint circuits. The constraint circuit represents relations between values of operands and inputs/output of the MUT. The experimental results showed that we could generate an accurate and compact constraint circuit. Hence, we can perform test generation and test program synthesis efficiently. We obtained high fault efficiency for each module.

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## References

- [1] S. M. Thatte and J. A. Abraham, "HTest generation for microprocessors," *IEEE Trans. on Computers*, Vol. C-29, No.6, pp. 429–44, 1981.
- [2] K. Batcher and C. Papachristou, "HInstruction randomization self test for processor cores," *Proc. 17th VLSI Test Symp.*, pp. 34–40, 1999.
- [3] W.-C. Lai, A. Krstic and K.-T. Cheng, "HTest program synthesis for path delay faults in microprocessor cores," *Proc. Int. Test Conf.*, pp. 1080–1089, 2000.
- [4] V. Singh, K. K. Saluja, M. Inoue and H. Fujiwara, "HDelay fault testing of processor cores in functional mode," *IEICE Trans. on Information and Systems*, Vol. E88-D, No.3, pp. 610–618, 2005.
- [5] L. Chen, S. Ravi, A. Raghunathan and S. Dey, "HA scalable software-based self-test methodology for programmable processors," *Proc. 40th ACM/IEEE Design Automation Conf.*, pp. 548–553, 2003.
- [6] K. Kambe, M. Inoue and H. Fujiwara, "HEfficient template generation for instruction-based self-test of processor cores," *Proc. 13th Asian Test Symp.*, pp. 152–157, 2004.
- [7] M. Inoue, K. Kambe, N. Hoashi and H. Fujiwara, "HInstruction-based self-test for sequential modules in processors," *IEEE 5th Workshop on RTL and High Level Testing (WRTL'04)*, pp. 109–114, 2004.
- [8] M. Vedula, J. A. Abraham, J. Bhadra and R. S. Tupuri, "HA hierarchical test generation approach using program slicing techniques on hardware description languages," *J. Electronic Testing: Theory and Applications*, Vol. 19, No. 2, pp. 149–160, 2003.
- [9] Z. Navabi, *VHDL analysis and modeling of digital systems*, McGraw-Hill, New York, 1997.